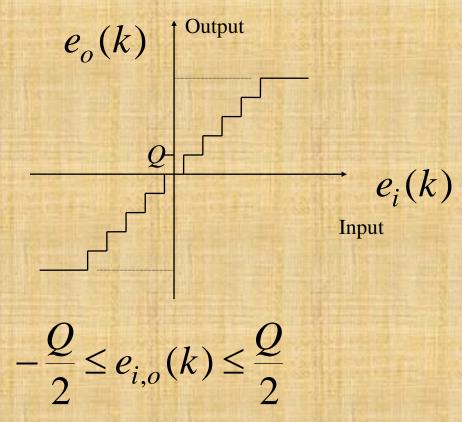
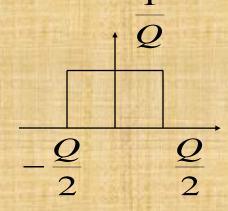
- Finite register lengths and A/D converters cause errors in:-
 - (i) Input quantisation.
 - (ii) Coefficient (or multiplier) quantisation
 - (iii) Products of multiplication truncated or rounded due to machine length

Quantisation



• The pdf for e using rounding



• Noise power
$$\sigma^2 = \int_{0/2}^{Q/2} e^2 p(e) . de = E\{e^2\}$$

or
$$\sigma^2 = \frac{Q^2}{12}$$

- Let input signal be sinusoidal of unity amplitude. Then total signal power $P = \frac{1}{2}$
- If b bits used for binary then $Q = 2/2^b$ so that $\sigma^2 = 2^{-2b}/3$
- Hence $P/\sigma^2 = \frac{3}{2}.2^{+2b}$

or
$$SNR = 1.8 + 6b$$
 dB

• Consider a simple example of finite precision on the coefficients a,b of second order system with poles $\rho e^{\pm j\theta}$

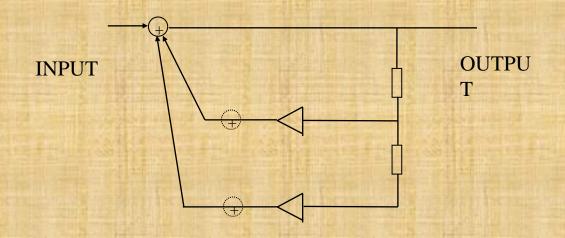
$$H(z) = \frac{1}{1 - az^{-1} + bz^{-2}}$$

$$H(z) = \frac{1}{1 - 2\rho\cos\theta \cdot z^{-1} + \rho^2 \cdot z^{-2}}$$

• where $a = 2\rho \cos \theta$ $b = \rho^2$

bit pattern	$2\rho\cos\theta, \rho^2$	ρ
000	0	0
001	0.125	0.354
010	0.25	0.5
011	0.375	0.611
100	0.5	0.707
101	0.625	0.791
110	0.75	0.866
111	0.875	0.935
1.0	1.0	1.0

• Finite wordlength computations





Limit-cycles; "Effective Pole" Model; Deadband

- Observe that for $H(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$
- instability occurs when $|b_2| \rightarrow 1$
- i.e. poles are
 - (i) either on unit circle when complex
 - (ii) or one real pole is outside unit circle.
- Instability under the "effective pole" model is considered as follows

- In the time domain with $H(z) = \frac{Y(z)}{X(z)}$
- $y(n) = x(n) b_1 y(n-1) b_2 y(n-2)$
- With $|b_2| \rightarrow 1$ for instability we have $Q[b_2y(n-2)]$ indistinguishable from y(n-2)
- Where Q[.] is quantisation

• With rounding, therefore we have

$$b_2 y(n-2) \pm 0.5$$
 $y(n-2)$ are indistinguishable (for integers)

or
$$b_2 y(n-2) \pm 0.5 = y(n-2)$$

• Hence
$$y(n-2) = \frac{\pm 0.5}{1-b_2}$$

With both positive and negative numbers

$$y(n-2) = \frac{\pm 0.5}{1 - |b_2|}$$

 $\frac{\pm 0.5}{1-|b_2|}$ The range of integers

constitutes a set of integers that cannot be individually distinguished as separate or from the asymptotic system behaviour.

• The band of integers
$$\left(-\frac{0.5}{1-|b_2|}, +\frac{0.5}{1-|b_2|}\right)$$

is known as the "deadband".

• In the second order system, under rounding, the output assumes a cyclic set of values of the deadband. This is a <u>limit-cycle</u>.

Consider the transfer function

$$G(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$$

$$y_k = x_k - b_1 y_{k-1} - b_2 y_{k-2}$$

• if poles are complex then impulse response is given by h_k

$$h_k = \frac{\rho^k}{\sin \theta} \cdot \sin \left[(k+1)\theta \right]$$

- Where $\rho = \sqrt{b_2}$ $\theta = \cos^{-1}\left(\frac{-b_1}{2\sqrt{b_2}}\right)$
- If $b_2 = 1$ then the response is sinusiodal with frequency

$$\omega = \frac{1}{T} \cos^{-1} \left(-\frac{b_1}{2} \right)$$

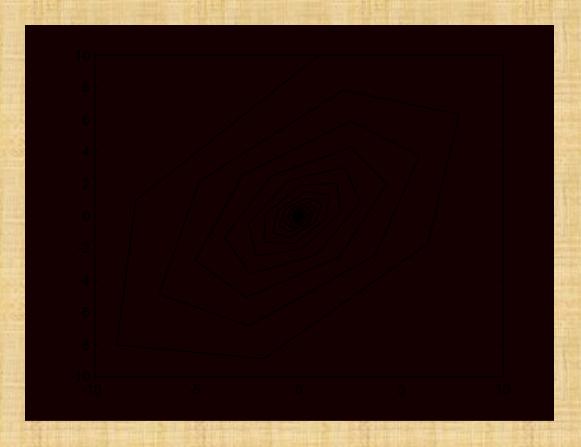
• Thus product quantisation causes instability implying an <u>"effective"</u> $b_2 = 1$.

Consider infinite precision computations for

$$y_k = x_k + y_{k-1} - 0.9 y_{k-2}$$

$$x_0 = 10$$

$$x_k = 0; k \neq 0$$



Now the same operation with integer precision



• Notice that with infinite precision the response converges to the origin

 With finite precision the reponse does not converge to the origin but assumes cyclically a set of values –the Limit Cycle

• Assume $\{e_1(k)\}$, $\{e_2(k)\}$ are not correlated, random processes etc.

$$\sigma_{0i}^{2} = \sigma_{e}^{2} \sum_{k=0}^{\infty} h_{i}^{2}(k) \quad \sigma_{e}^{2} = \frac{Q^{2}}{12}$$
Hence total output noise power

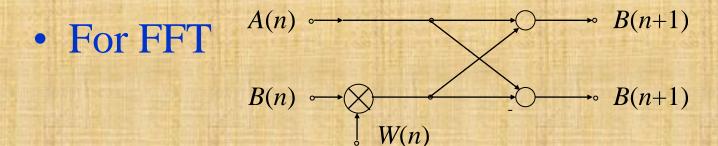
$$\sigma_0^2 = \sigma_{01}^2 + \sigma_{02}^2 = 2 \cdot \frac{2^{-2b}}{12} \sum_{k=0}^{\infty} \rho^{2k} \cdot \frac{\sin^2[(k+1)\theta]}{\sin^2 \theta}$$

• Where $Q = 2^{-b}$ and

$$h_1(k) = h_2(k) = \rho^k \cdot \frac{\sin[(k+1)\theta]}{\sin \theta}; \ k \ge 0$$

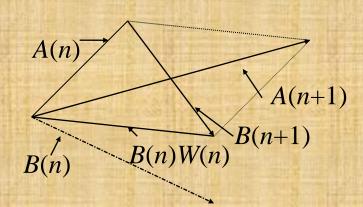
• ie

$$\sigma_0^2 = \frac{2^{-2b}}{6} \left[\frac{1+\rho^2}{1-\rho^2} \cdot \frac{1}{1+\rho^4 - 2\rho^2 \cos 2\theta} \right]$$



$$A(n + 1) = A(n) + W(n).B(n)$$

 $B(n + 1) = A(n) - W(n).B(n)$



• FFT

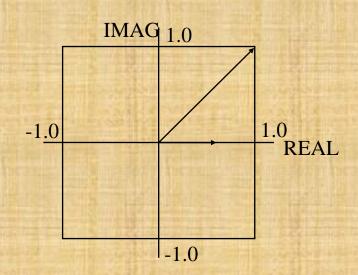
$$|A(n+1)|^{2} + |B(n+1)|^{2} = 2$$

$$|A(n+1)|^{2} = 2|A(n)|^{2}$$

$$|A(n)| = \sqrt{2}|A(n)|$$

AVERAGE GROWTH: 1/2 BIT/PASS

• FFT



$$A_{x}(n+1) = A_{x}(n) + B_{x}(n)C(n) - B_{y}(n)S(n)$$

$$|A_{x}(n+1)| < |A_{x}(n)| + |B_{x}(n)|C(n)| - |B_{y}(n)|S(n)|$$

$$\frac{|A_{x}(n+1)|}{|A_{x}(n)|} < 1.0 + |C(n)| - |S(n)| = 2.414....$$

• PEAK GROWTH: 1.21.. BITS/PASS

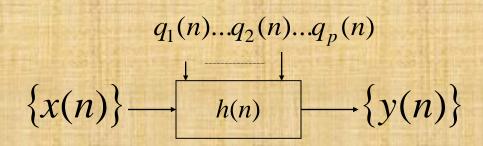
Linear modelling of product quantisation

$$\xrightarrow{x(n)} Q[\cdot] \xrightarrow{\widetilde{x}(n)}$$

Modelled as

$$x(n) \longrightarrow \widetilde{x}(n) = x(n) + q(n)$$
 $q(n)$

- For <u>rounding</u> operations q(n) is uniform distributed between $-\frac{Q}{2}$, $\frac{Q}{2}$ and where Q is the quantisation step (i.e. in a wordlength of bits with sign magnitude representation or mod 2, $Q = 2^{-b}$).
- A discrete-time system with quantisation at the output of each multiplier may be considered as a multi-input linear system



Then

$$y(n) = \sum_{r=0}^{\infty} x(r).h(n-r) + \sum_{\lambda=1}^{p} \left[\sum_{r=0}^{\infty} q_{\lambda}(r).h_{\lambda}(n-r) \right]$$

• where $h_{\lambda}(n)$ is the impulse response of the system from λ the <u>output</u> of the multiplier to y(n).

• For zero input i.e. x(n) = 0, $\forall n$ we can write

$$|y(n)| \le \sum_{\lambda=1}^{p} |\hat{q}_{\lambda}| \cdot \sum_{r=0}^{\infty} |h_{\lambda}(n-r)|$$

• where $|\hat{q}_{\lambda}|$ is the maximum of $|q_{\lambda}(r)|$, $\forall \lambda, r$ which is not more than \underline{Q}

• ie
$$|y(n)| \le \frac{Q}{2} \cdot \sum_{\lambda=1}^{p} \left[\sum_{n=0}^{\infty} |h_{\lambda}(n-r)| \right]$$

However

$$\sum_{n=0}^{\infty} |h_{\lambda}(n)| \le \sum_{n=0}^{\infty} |h(n)|$$

And hence

$$|y(n)| \le \frac{pQ}{2} \cdot \sum_{n=0}^{\infty} |h(n)|$$

• ie we can estimate the maximum swing at the output from the system parameters and quantisation level